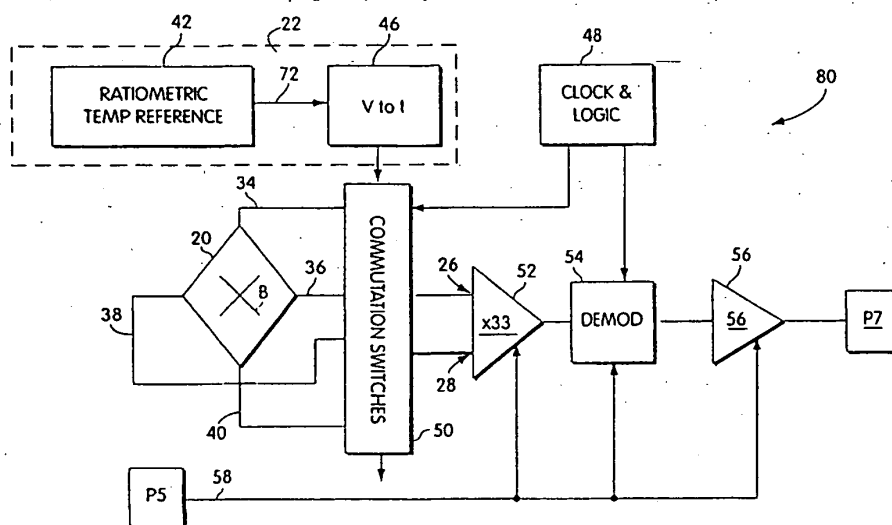




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(54) Title: MONOLITHIC MAGNETIC SENSOR HAVING EXTERNALLY ADJUSTABLE TEMPERATURE COMPENSATION



(57) Abstract

The current flow from a temperature-variable current source to a Hall-effect element is adjusted according to sensed temperature conditions of the element to compensate for temperature-dependent changes in the magnetic-field sensitivity of the Hall-effect element and in the magnitude of the permanent magnetic fields of magnetic components sensed by the element. A trimmable resistor is connected between two external terminals of a Monolithically integrated circuit to provide external control over the sensitivity of the temperature variable current source to changing temperature conditions. The device also alternately switches the quadrature states of output and bias supply contacts of the Hall-effect element to compensate for the offset and drift thereof.

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**MONOLITHIC MAGNETIC SENSOR HAVING EXTERNALLY ADJUSTABLE
TEMPERATURE COMPENSATION**

Background of The Invention

5 **1. Field of the Invention**

The present invention relates to magnetic sensors and, more particularly, to a Hall-effect sensor arrangement having temperature compensation circuitry that is user-adjustable.

2. Discussion of Related Art

10 Hall-effect elements are used to sense magnetic fields. In its simplest form, a monolithically fabricated Hall-effect element employs a square plate formed, for example, of an epitaxial (epi) pocket formed in a bipolar (BP) or bipolar complementary metal oxide semiconductor (BiCMOS) process. The square plate has two pairs of contacts oriented such that each pair of contacts straddles the plate from a different pair of its diametrically opposing corners. When a current I_H is caused to flow between one of the two pairs of contacts, a
15 voltage V_H will appear across the other pair of contacts. The voltage V_H is a function of: (a) the amplitude of the current I_H , (b) the strength of a magnetic field B , if any, intercepting the Hall plate, and (c) the sensitivity S_H of the Hall plate to the magnetic field B (i.e., $V_H = I_H * B * S_H$, where $*$ denotes the multiplication operator). Thus, when properly calibrated, a Hall
20 plate may be used to measure the strength of a magnetic field B by applying a known current I_H between one pair of contacts of a Hall plate and measuring the voltage V_H across the other pair of contacts.

25 There are two major difficulties encountered when using Hall plates to sense magnetic fields. One difficulty arises because a Hall plate includes unavoidable imperfections, such as geometrical asymmetries, that cause the Hall plate to produce a non-zero output voltage V_H in the absence of any applied magnetic field B . This voltage V_H generated when $B=0$ commonly is referred to as the offset voltage V_{OH} of the Hall plate. A discussion of the Hall effect, the operation of Hall-effect elements, circuits in which Hall-effect elements may be used, and various techniques for addressing the problem of Hall plate offset voltage are discussed in the following patents and publications, each of which is incorporated herein by
30 reference: Bilotti et al., U.S. Patent 5,621,319; Mehrgardt et al., U.S. Patent 5,406,202; Bilotti, Albert, "Monolithic Magnetic Hall Sensor Using Dynamic Quadrature Offset Cancellation," IEEE Journal of Solid-State Circuits, Vol. 32, No. 6, June 1997; Bellekom, A.A., and Munter,

P.J.A., "Offset Reduction in Spinning-Current Hall Plates," Sensors and Materials, 5, 5 (1994) 253-263, MYU Tokyo; Baltes, H.P., and Popovic, R.S., "Integrated Semiconductor Magnetic Field Sensors," Proc. IEEE, vol. 74, pp. 1107-1132, Aug. 1986.

A second difficulty arises due to the fact that the sensitivity S_H of a Hall-effect element changes as the temperature of the element changes. Therefore, for a given current I_H and magnetic field B , the voltage V_H produced by a Hall-effect element will vary depending on the temperature of the element. Because of this temperature-dependency, the actual magnitude of the magnetic field B is difficult to ascertain when a Hall-effect element is used in an environment wherein its temperature is not held constant.

In addition, Hall-effect elements often are used to sense the permanent magnetic fields of components. For example, a Hall-effect element may be used to sense the position of a magnetic component while it is rotating or otherwise in motion relative to the Hall-effect element. Because the magnitude of a magnetic field emanating from a magnetic component changes as the temperature of the component changes, for any given position of the component with respect to the Hall-effect element, the magnitude of the output voltage V_H of the Hall-effect element when the component is at a first temperature will not be the same as the output voltage V_H when the component is at a second temperature. Therefore, the magnitude of the output voltage V_H of a Hall-effect element does not provide a reliable indication of the position of a magnetic component when the component is subjected to a variable-temperature environment.

It therefore is a general object of the present invention to provide a Hall effect device having improved immunity to the effects of a variable-temperature environment.

Summary of the Invention

According to one aspect of the present invention, a Hall effect device includes a substrate mounted within an integrated circuit package, a Hall-effect element integrated on the substrate, and a pair of terminals adapted to interface with a component external to the package. The pair of terminals is arranged with respect to the Hall-effect element such that, when the component is coupled to the pair of terminals, a characteristic of the component may be changed to adjust the sensitivity of the gain of the Hall effect device to temperature changes in the device.

According to another aspect, the Hall effect device includes a temperature-variable

signal source arranged to provide the bias signal such that a value of the bias signal changes in response to temperature changes in the Hall-effect element. Additionally, the pair of terminals is arranged with respect to the signal source such that, when the component is coupled to the pair of terminals, the characteristic of the component may be changed to adjust the sensitivity of the signal source to temperature changes in the Hall-effect element.

According to another aspect of the invention, a Hall-effect device includes an integrated circuit package having a Hall-effect element mounted within it, and a user-accessible component arranged with respect to the Hall-effect element such that a characteristic of the component may be changed to adjust the sensitivity of the gain of the Hall-effect device to temperature changes in the device.

According to another aspect a Hall-effect device includes a Hall effect device mounted in an integrated circuit package and means for adjusting the sensitivity of the gain of the Hall-effect device to changing temperature conditions from a point external to the package.

According to yet another aspect of the invention, a method for compensating for the effects of temperature variations on a Hall-effect element includes coupling a component to a pair of external terminals of an integrated circuit to adjust the sensitivity of the gain of a Hall-effect device included in the integrated circuit.

Brief Description of the Drawings

Figure 1 is a top-view of a quad-cell Hall-effect element used in one embodiment of the invention.

Figure 2 is a side cross-sectional view of the substrate portion of the quad-cell Hall cell shown in Figure 1.

Figure 3 is a block diagram of one embodiment of a Hall-effect device according to the invention.

Figure 4 is a partial schematic/partial block diagram showing in more detail a commutation switch portion of the device shown in Figure 3.

Figure 5 is a partial schematic/partial block diagram showing in more detail a demodulation portion of the device shown in Figure 3.

Figure 6 is a timing diagram showing control and output signals of the portions of the device shown in Figures 4 and 5.

Figure 7 is a schematic diagram showing in more detail the ratiometric temperature

reference portion of the circuit shown in Figure 3.

Figure 8 is a graph showing how voltages produced by the temperature reference portion shown in Figure 7, which are presented on external pins of the device, vary with temperature.

5 Figure 9 is partial block/partial schematic diagram showing a possible configuration of circuit elements external to the device.

Figure 10 is a side-view of an embodiment of the device according to the invention showing how the device senses a magnetic field applied thereto.

10 Figure 11 is partial block/partial schematic diagram showing a possible configuration of circuit elements external to the device.

Detailed Description of the Invention

Figure 1 is a top-view of a quad-cell Hall-effect element 20 used in one embodiment of the invention. As shown, quad-cell Hall-effect element 20 includes four Hall cells 20A-
15 20D. Hall cells 20A, 20B, 20C and 20D are made of n-type epitaxial (n-epi) regions (e.g., 200 micrometers in diameter) formed within p-type wells or hubs (p-wells/p-tubs) 30A, 30B, 30C and 30D, respectively. P-wells/p-tubs 30A, 30B, 30C and 30D, in turn, are formed on an n-type substrate 32.

Each of Hall cells 20A, 20B, 20C and 20D includes two pairs of orthogonally oriented
20 contacts. Specifically, Hall cells 20A, 20B, 20C and 20D, include, respectively, first pairs of contacts 20A1 and 20A3, 20B1 and 20B3, 20C1 and 20C3, and 20D1 and 20D3, and second pairs of contacts 20A2 and 20A4, 20B2 and 20B4, 20C2 and 20C4, and 20D2 and 20D4. Conductor 34 interconnects contacts 20A1, 20B2, 20C3 and 20D4, conductor 36 interconnects contacts 20A2, 20B3, 20C4 and 20D1, conductor 38 interconnects contacts
25 20A4, 20B1, 20C2 and 20D3, and conductor 40 interconnects contacts 20A3, 20B4, 20C1 and 20D2, thereby wiring cells 20A-20D in parallel such that each of cells 20A-20D is oriented orthogonally with respect to the other cells in quad-cell Hall-effect element 20. This orientation of cells 20A-20D reduces the adverse effects of imperfections, such as processing gradients, in the cells.

30 Figure 1 also shows how each of cells 20A-20D may be intercepted by a magnetic field B , which is shown as entering the surface of the cells in a direction normal and into the page. If a current I_H were applied between conductors 34 and 40, for example, then the

magnitude of magnetic field B and the sensitivity S_H of the Hall-effect element would determine a magnitude of a voltage V_H created between conductors 36 and 38 according to the equation $V_H = I_H * B * S_H$, discussed above.

Figure 2 is a side cross-sectional view of the substrate portion of the quad-cell Hall-effect element 20 shown in Figure 1. As shown, p-wells/p-tubs 30D and 30C are formed within an n-type substrate 32, and n-EPI Hall cells 20C and 20D are formed within p-wells/p-tubs 30C and 30D, respectively. The letters "p" and "n" in Figures 1 and 2 are intended only to illustrate the doping types of the various regions, and are not intended to signify the amounts by which the regions are doped or the process by which the regions were formed.

Figure 3 is a block diagram of one embodiment of a Hall-effect device 80 according to the invention. As shown, Hall-effect device 80 includes quad-cell Hall-effect element 20, commutation switches 50, a differential pre-amplifier 52, a demodulator 54, an output amplifier 56, an output terminal P7, a reference terminal P5, a temperature-variable current source 22 (which includes a ratiometric temperature reference block 42 and a voltage-to-current converter 46), and a clock and logic stage 48.

The operation of circuit 80 is discussed in more detail below in connection with a description of its various portions, but Figure 3 is useful to understand the basic operation of Hall-effect device 80. As shown, temperature-variable current source 22 supplies a Hall bias current I_H which varies according to the temperature of the device to commutation switches 50. Commutation switches 50 are controlled by clock and logic stage 48 to commute the current I_H alternately (1) between conductors 34 and 40, and (2) between conductors 36 and 38. When the current I_H is presented between conductors 34 and 40, commutation switches 50 concurrently apply the voltage generated between conductors 36 and 38 (responsive to the magnetic field B) to inputs 26 and 28, respectively, of differential pre-amplifier 52; and when the current I_H is presented between conductors 36 and 38, commutation switches 50 concurrently apply the voltage generated between conductors 34 and 40 (responsive to the magnetic field B) to inputs 26 and 28, respectively, of differential pre-amplifier 52. This commutation of signals to and from quad-cell Hall-effect element 20 is performed to eliminate the inherent voltage offset V_{OH} of the Hall cells, and is discussed in detail in several of the documents referred to and incorporated by reference above. Suffice it to note, however, that the flowing of current in a first direction across a Hall cell generates a first voltage V_1 including a first Hall voltage V_{H1} and a first Hall offset voltage V_{OH1} ($V_1 = V_{H1} + V_{OH1}$), and the

flowing of current across a Hall cell in a second direction, which is oriented orthogonally to the first direction, generates a second voltage V_2 including a second Hall voltage V_{H2} and a second Hall offset voltage V_{OH2} ($V_2 = -V_{H2} + V_{OH2}$).

Because the second hall voltage V_{H2} is approximately equal to and of an opposite sign as the first Hall voltage V_{H1} (i.e., $V_{H2} \approx -V_{H1} = -V_H$) and the second Hall offset voltage V_{OH2} is approximately equal to and the same sign as the first Hall offset voltage V_{OH1} (i.e., $V_{OH2} \approx V_{OH1} = V_{OH}$), a subtraction of the voltage V_2 generated by the current flowing in the second direction from the voltage V_1 generated by the current flowing in the first direction (i.e., $(V_{H1} + V_{OH1}) - (-V_{H2} + V_{OH2})$) effectively cancels the offset voltage so that only a voltage equal to $2V_H$ results. As explained below, this subtraction is performed by a combination of pre-amplifier 52 and demodulator 54. The offset-free output of demodulator 54 then is provided to the input of rail-to-rail output amplifier 56, and the output of amplifier 56 feeds external terminal P7 (e.g., a pin) of the device. As described below, the gain of amplifier 56 is adjusted using resistors external to device 80.

The effect of temperature on the sensitivity of quad-cell Hall-effect element 20 is compensated by adjusting the magnitude of the Hall bias current I_H supplied to commutation switches 50. That is, because the voltage V_H produced at the output of Hall-effect element 20 is proportional to the magnitude of the current flowing through the element as well as to the sensitivity constant S_H (i.e., $V_H = I_H \cdot B \cdot S_H$), an increase or decrease in sensitivity S_H due to a temperature change may be countered by a corresponding decrease or increase, respectively, in Hall current I_H . The output voltage V_H therefore always will be the same for a given magnitude of magnetic field B , regardless of the temperature of the Hall-effect device 80.

Similarly, an increase or decrease in the magnetic field strength B of a magnetic component being sensed by device 80 due to a temperature change may be countered by a corresponding decrease or increase, respectively, in the Hall current I_H . This is useful when the magnetic field is produced by a moving permanent magnet. Thus, regardless of the temperature of the environment in which the magnetic component and Hall-effect device 80 are disposed, the output voltage V_H of Hall-effect device 80 always will be the same when the moving (e.g., rotating) component being sensed is a given distance from device 80.

Figure 4 is a partial schematic/partial block diagram showing commutation switches 50 and their surrounding circuitry in more detail. As shown, commutation switches 50 include switches S1-S8. Switches S1 and S2 are connected between a first polarity output of

temperature-variable current source 22 and conductors 36 and 34, respectively, of quad-cell Hall-effect element 20, and switches S3 and S4 are connected between a second polarity output of temperature-variable current source 22 and conductors 38 and 40, respectively, of Hall-effect element 20. Similarly, switches S5 and S6 are connected, respectively, between
5 conductors 34 and 36 of Hall-effect element 20 and non-inverting input 26 of pre-amplifier 52, and switches S7 and S8 are connected, respectively, between conductors 40 and 38 of Hall-effect element 20 and inverting input 28 of pre-amplifier 52. As explained below, the symbol "P1" or "P2" next to switches S1-S8 indicates which of two phases of a control signal from clock and logic stage 48 (shown in Figure 3) causes each of the switches to close. In
10 addition to receiving signals at non-inverting input 26 and inverting input 28, pre-amplifier 52 receives a reference voltage V_{REF} from reference conductor 58. The reference voltage V_{REF} is created within Hall-effect device 80 and is set to be one-half of the high-supply voltage VCC of device 80 (i.e., $V_{REF}=VCC/2$). The high-supply voltage VCC is approximately 5 volts, so the reference voltage V_{REF} is set to approximately 2.5 volts. The
15 operation of the circuitry shown in Figure 4 is described below in connection with the description of the timing diagram shown in Figure 6.

Figure 5 is a partial schematic/partial block diagram showing demodulation block 54 in more detail. As shown, demodulation block 54 includes capacitors C1 and C2 and switches S9 and S10. Capacitor C1 is connected between a single-ended output of differential pre-
20 amplifier 52 and one terminal of each of switches S9 and S10. The other terminal of switch S9 is connected to reference conductor 58 (which is maintained at V_{REF}), and the other terminal of switch S10 is connected to a non-inverting input of output amplifier 56. Capacitor C2 is connected between the non-inverting input of amplifier 56 and conductor 58.

Figure 6 is a timing diagram showing control signals for switches S1-S10 and output
25 signals of pre-amplifier 52 and demodulator 54 shown in Figures 4 and 5. Specifically, the timing diagram of Figure 6 illustrates: (1) commutation clock signal 60, (2) pre-amplifier output signal 62, (3) switch S10 control signal 64, (4) switch S9 control signal 66, and (5) demodulator output signal 68. The vertical axis of each of these signals represents an amplitude attained by the signal and the horizontal axis represents time. While all of the
30 signals shown in Figure 6 share a common time axis, the placement of one signal above another is not intended to indicate that the signal attains a higher amplitude than do the others.

Referring now to Figure 4 in conjunction with Figure 6, the operation of the portion of

Hall-effect device 80 shown in Figure 4 will be explained. As shown, commutation clock 60 has two phases, P1 and P2. Specifically, when clock 60 is low, it is in phase P1, and when clock 60 is high, it is in phase P2. Each of switches S1-S8 is arranged to be closed during only one of these two phases. That is, switches S2, S4, S6 and S8 are closed during phase P1 (when clock 60 is low) and switches S1, S3, S5 and S7 are closed during phase P2 (when clock 60 is high). The phase during which each of the switches is closed is indicated next to the switch by the symbol "P1" or "P2."

During phase P1 of clock 60, temperature-variable current source 22 is connected between conductors 34 and 40 of Hall-effect element 20 and conductors 36 and 38 of Hall-effect element 20 are connected, respectively, to non-inverting input 26 and inverting input 28 of differential pre-amplifier 52. Pre-amplifier 52 amplifies the differential voltage between its non-inverting input 26 and its inverting input 28, and effectively adds the amplified input voltage (which is positive during phase P1) to the reference voltage V_{REF} on conductor 58. Pre-amplifier 52 then provides this sum of voltages as a single-ended output signal (shown as pre-amplifier output signal 62 in Figure 6). As shown, pre-amplifier output signal 62 is greater than V_{REF} during phase P1 (e.g., as indicated by portion 62A of curve 62).

During phase P2 of clock 60, temperature-variable current source 22 is connected between conductors 36 and 38 of Hall-effect element 20 and conductors 34 and 40 of Hall-effect element 20 are connected, respectively, to non-inverting input 26 and inverting input 28 of differential pre-amplifier 52. Pre-amplifier 52 amplifies the differential voltage between its non-inverting input 26 and its inverting input 28, and effectively adds the amplified input voltage (which is negative during phase P2) to the reference voltage V_{REF} on conductor 58. Pre-amplifier 52 then provides this sum of voltages as a single-ended output signal (shown as pre-amplifier output signal 62 in Figure 6). As shown, pre-amplifier output signal 62 is less than V_{REF} during phase P2 (e.g., as indicated by portion 62B of curve 62).

Generally, due to the offset voltage V_{OH} of Hall-effect element 20, portions 62A and 62B, respectively, of curve 62 will not be the same voltage greater than (V_H) and less than ($-V_H$) the voltage V_{REF} during phases P1 and P2. Demodulator 54 deals with this difference by effectively subtracting the voltage $-V_H$ at the output of pre-amplifier 52 during phase P2 from the voltage V_H at the output of pre-amplifier 52 during phase P1, thereby removing the effects of the Hall offset voltage V_{OH} , as described above.

Referring now to Figure 5 in conjunction with Figure 6, the operation of demodulator

54 will be described. Clock and logic stage 48 (shown in Figure 3) produces switch S9 control signal 66 and switch S10 control signal 64 as shown in Figure 6. Switches S9 and S10 are arranged such that they are closed, respectively, during phase P4 of switch S9 control signal 66 and during phase P3 of switch S10 control signal 64 (i.e., when switch S9 control signal 66 and switch S10 control signal 64, respectively, are high). The phase during which each of switches S9 and S10 is closed is indicated next to the switch by the symbol "P4" or "P3."

During phase P4, switch S9 is closed during a brief time interval while the output of pre-amplifier 52 is negative with respect to V_{REF} so that capacitor C1 will charge to the current voltage at the output of pre-amplifier 52 with respect to the voltage V_{REF} . During phase P3, after switch S9 has opened, switch S10 is closed during a brief time interval while the output of pre-amplifier 52 is positive with respect to V_{REF} . Because capacitors C1 and C2 are connected in series during this interval, some of the charge on capacitor C1 will be transferred to capacitor C2. After several P4/P3 cycles, the charge on capacitor C2 will be equal to the peak-to-peak voltage of curve 62.

In this manner, demodulator circuit 54 effectively subtracts (a) the negative difference between the amplitude of the voltage at the output of pre-amplifier 52 and reference voltage V_{REF} during phase P2 from (b) the positive difference between the amplitude of pre-amplifier 52 and reference voltage V_{REF} during phase P1. The output of demodulator circuit 54, i.e., the pole of capacitor C2 that is connected to the non-inverting input of output amplifier 56, therefore remains at a positive voltage equal to the peak-to-peak voltage of curve 62. As discussed above, this peak-to-peak voltage has the Hall offset voltage V_{OH} removed from it.

Figure 7 is a schematic diagram showing ratiometric temperature reference block 42 in more detail. As shown, ratiometric temperature reference stage 42 includes operational amplifiers 74 and 86, temperature-sensitive n+ diffusion resistor DR1, temperature-sensitive p-well diffusion resistors DR2 and DR3, and thin film resistors TF1-TF7, which are not particularly sensitive to temperature changes.

A resistive bridge including diffusion resistor DR1 and thin-film resistors TF1 and TF2 is connected between high-supply rail 70 (which has a voltage V_{CC} thereon) and an analog ground 84. specifically, thin film resistor TF1 is connected between high-supply rail 70 and a non-inverting input of operational amplifier 74, and resistors DR1 and TF2 are connected in series between the non-inverting input of operational amplifier 74 and analog

ground 84. The non-inverting input of operational amplifier 74 therefore is maintained at a voltage that changes in proportion to changes in the supply voltage (i.e., it is a ratiometric differential reference voltage).

Resistors TF3 and TF4 are connected in series between high-supply rail 70 and analog ground 84, and the connection point of these resistors is connected to the inverting input of operational amplifier 74. Resistor TF5 is connected in feedback between the output and the inverting input of operational amplifier 74. The values of resistors TF3-TF5 therefore control the gain of operational amplifier 74.

The output of operational amplifier 74 feeds node 88. Node 88, in turn, is connected to: (1) an external terminal P1, (2) one terminal of a first divider resistor R1 (including series-connected resistors DR2 and TF6), the other terminal of which is connected to an external terminal P2, and (3) one input of unity-gain inverting operational amplifier 86, the other input of which is connected to conductor 58 (which has the voltage V_{REF} applied on it). Thus, operational amplifier 86 will produce an output that is equal to the inverted difference between: (a) the voltage at node 88, and (b) the reference voltage $VCC/2$. As used herein, an "external terminal" refers to a terminal extending outwardly from an integrated circuit package such the terminal may interface with circuitry external to the package.

The output of operational amplifier 86 is connected to one terminal of a second divider resistor R2 (including series-connected resistors DR3 and TF7), the other terminal of which is connected to: (a) an external terminal P3, and (b) conductor 72, which drives a control input of voltage-to-current converter 46 (shown in Figure 3). Voltage-to-current converter 46 is configured so that a one millivolt change in the voltage applied to its control input results in a 0.04% change in Hall bias current I_H provided to quad-cell Hall-effect element 20 (shown in Figure 3).

Additionally, a trimmable resistor R3 is selectively connected between external terminal P3 and one of external terminals P1 and P2. Resistor R3 preferably is external to the integrated circuit package in which Hall-effect device 80 is disposed so that its value is user-selectable. Resistor R3 may comprise a simple potentiometer, a thick-film laser trimmable resistor, or any other device having a user-adjustable resistance known to those skilled in the art. It should be noted, however, that resistor R3 also may be included within the same package as Hall-effect device 80 so long as it is readily accessible to an end user of the integrated circuit.

As noted above, the resistance of diffusion resistors DR1, DR2 and DR3 changes responsive to temperature changes. Thus, as the temperature of Hall-effect device 80 increases, the resistance of resistor DR1 will change and the voltage at the output of operational amplifier 74 (node 88) will change accordingly. Assuming initially that resistor R3 is not present, when the voltage at node 88 is equal to $VCC/2$ (i.e., the voltage on conductor 58), the voltage at each of external terminals P1, P2 and P3 also will be equal to $VCC/2$. As the voltage at node 88 diverges from $VCC/2$, however, the voltage at external terminals P1 and P2 will follow the divergence of the voltage at node 88, and the presence of polarity-inverting operational amplifier 86 will cause the voltage at external terminal P3 to diverge in a voltage direction opposite the direction in which the voltage at node 88 diverges from $VCC/2$. Specifically, the voltage at external terminal P3 will diverge in a positive voltage direction when the voltage at node 88 decreases below $VCC/2$ and will diverge in a negative voltage direction when the voltage at node 88 increases above $VCC/2$.

After fabrication of the device, bridge resistors TF1 and TF2, or gain-setting resistors TF3-TF5, may be trimmed using any conventional wafer probing apparatus so that the voltage at node 88 is equal to $VCC/2$ when the temperature of the Hall-effect device is at an ambient temperature (e.g., approximately 27° Celsius).

Figure 8 is a graph showing how the voltages on external terminals P1, P2 and P3 vary with temperature, both with and without resistor R3 being connected between terminal P3 and one of terminals P1 and P2. As shown, curve 76 represents the voltage on both of external terminals P1 and P2 when resistor R3 is not connected to either of them, curve 78 represents the voltage on external terminal P3 when resistor R3 is not connected to it, and curve 82 represents the voltage on external terminal P3 when resistor R3 is connected between terminal P3 and one of terminals P1 and P2. Curves 76, 78 and 80 are shown as they would appear after bridge resistors TF1 and TF2, or gain-setting resistors TF3-TF5, have been trimmed properly. That is, after selected ones of resistors TF1-TF5 have been trimmed properly, curves 76, 78 and 80 will converge at a point 80, which corresponds to an ambient temperature of 27° Celsius.

When resistor R3 is connected between terminal P3 and one of terminals P1 and P2, external terminal P3 will form a node of a voltage divider including at least two resistors that divide the voltage between node 88 and the output of operational amplifier 86. Specifically, when resistor R3 is connected between terminal P3 and terminal P1, terminal P3 will form a

node of a voltage divider including resistor R2 (which includes resistors DR3 and TF7) and resistor R3, and when resistor R3 is connected between terminal P3 and terminal P2, terminal P3 will form a node of a voltage divider including resistor R1 (which includes resistors DR2 and TF6), resistor R2 (which includes resistors DR3 and TF7), and resistor R3. The voltage at terminal P3 resulting from the voltage division performed by these resistors is represented by curve 82 in Figure 8. The ability to connect resistor R3 to either of two separate terminals is provided to permit temperature compensation of the device for different types of magnets. For example, the connection of resistor R3 between terminals P1 and P3 may permit the compensation of a -2000 parts-per-million (ppm) magnet and the connection of resistor R3 between terminals P2 and P3 may permit the compensation of a -200 ppm magnet.

After resistor R3 is connected between external terminal P3 and one of external terminals P1 and P2, its value may be trimmed to optimize the slope of curve 82 for the particular application. That is, by adjusting the slope of curve 82, the amount that the voltage at external terminal P3 changes in response to temperature changes will be adjusted. Because the voltage at external terminal P3 controls the magnitude of the Hall current I_H generated by voltage-to-current converter 42 (shown in Figure 3), optimizing the slope of curve 82 (by trimming resistor R3) will permit compensation of the temperature dependency of the sensitivity S_H of Hall-effect element 20 with a high degree of accuracy.

As mentioned above, each of resistors R1 and R2 includes a series combination of a thin film and a temperature-sensitive diffusion resistor. Diffusion resistors are used in resistors R1 and R2 to add curvature to curve 82 to provide correction for the second-order temperature coefficient of the quad-cell Hall-effect element sensitivity S_H .

Figure 9 is partial block/partial schematic diagram showing a possible embodiment of Hall-effect device 80 and a possible configuration of circuit elements external thereto. In the embodiment shown, Hall-effect device 80 is an 8-pin dual-in-line package (DIP) including terminals (i.e., external connectors) P1-P8, several of which are the same terminals shown in previously-described Figures. Terminal P8 receives power from high-supply conductor 70 and terminal P4 is connected to an analog ground conductor 84. Trimmable resistor R3 is connected between terminals P2 and P3 (or optionally could be connected between terminals P1 and P3). Resistors R4-R6 are configured to set the gain of output amplifier 56. Specifically, resistor R4 is connected between terminals P8 and P6 (which is the inverting input of output amplifier 56), resistor R6 is connected between terminals P6 and P5 (which

has a voltage equal to $VCC/2$ on it), and resistor R5 is connected between terminals P6 and P7 (which is the output of amplifier 56).

Figure 10 is a side-view of an embodiment of Hall-effect device 80 showing how device 80 senses a magnetic field. As shown, the device is configured to sense a magnetic
5 field **B** oriented transverse to the plane of the device.

Figure 11 is partial block/partial schematic diagram showing another possible configuration of circuit elements external to device 80, wherein an additional potential divider (including resistors R7 and R10) is connected across the supply to further adjust the offset of the device. Specifically, resistor R7 is connected between terminal P5 and a first terminal of
10 resistor R9 (the second terminal of which is connected to terminal P7), resistor R8 is connected between terminal P6 and terminal P7, and resistor R10 is connected between the first terminal of resistor R9 and terminal P3.

While the embodiment of Hall-effect device 80 has been described herein as using, temperature sensitive resistors, any devices capable of producing an output that varies with
15 temperature could equivalently be used without departing from the intended scope of the invention. Additionally, while the temperature-variable current source 22 includes a first device that generates a voltage dependent on temperature and a second device that converts the generated voltage into a current to be supplied to Hall-effect element 20, a single device that produces a temperature-dependent current could equivalently be used, or, alternatively, a
20 temperature-dependent voltage source could be used directly as the source of current across terminal pairs of Hall-effect element 20. Further, while the external device used to adjust the sensitivity of the temperature compensation circuitry has been described herein as a resistor, any other external device capable of adjusting the sensitivity of the device equivalently could be employed. In addition, while the semiconductor components have been described herein as
25 being of particular doping types, opposite doping types equivalently may be used.

Having thus described at least one illustrative embodiment of the invention, various alterations, modifications and improvements will readily occur to those skilled in the art. Such alterations, modifications and improvements are intended to be within the spirit and scope of the invention. Accordingly, the foregoing description is by way of example only and
30 is not intended as limiting. The invention is limited only as defined in the following claims and the equivalents thereto.

What is claimed is:

CLAIMS

1. A Hall-effect device, comprising:

5 a substrate mounted in an integrated circuit package;
a Hall-effect element, integrated on the substrate, having at least one input for receiving a bias signal and at least one output for providing an output signal dependent on the bias signal and a magnetic field intercepting the Hall-effect element; and
a pair of terminals, adapted to have a component external to the package coupled
10 thereto, arranged with respect to the Hall-effect element such that, when the component is coupled to the pair of terminals, a characteristic of the component may be changed to adjust a sensitivity of a gain of the Hall-effect device to temperature changes in the device.

2. The Hall-effect device as claimed in claim 1, wherein:

15 the device further comprises a temperature-variable signal source arranged to provide the bias signal such that a value of the bias signal changes in response to temperature changes in the Hall-effect element, and

the pair of terminals is arranged with respect to the signal source such that, when the component is coupled to the pair of terminals, the characteristic of the component may be
20 changed to adjust a sensitivity of the signal source to temperature changes in the Hall-effect element.

3. The Hall-effect device as claimed in claim 2, wherein the signal source includes:

25 a controllable current source having a control input, and
a temperature-sensitive signal source integrated on the substrate and coupled to the current source to provide a temperature dependent signal to the control input of the controllable current source.

30 4. The Hall-effect device as claimed in claim 3, wherein:

the controllable current source includes a voltage-controlled current source, and
the temperature-sensitive signal source includes a temperature-sensitive voltage

source.

5 5. The Hall-effect device as claimed in claim 4, wherein:
 at least one of the pair of terminals is arranged to form a node of a potential divider
 when the component external to the package is coupled therebetween.

6. The Hall-effect device as claimed in claim 5, further comprising the component
 external to the package, which comprises a resistor having an adjustable resistance.

10 7. The Hall-effect device as claimed in claim 1, further comprising the component
 external to the package, which comprises a resistor having an adjustable resistance.

8. The Hall-effect device as claimed in claim 1, further comprising means for
 compensating for offset of the Hall-effect element.

15

9. A Hall-effect device, comprising:
 an integrated circuit package;
 a Hall-effect element, mounted in the package, having at least one input for receiving a
 bias signal and at least one output for providing an output signal dependent on the bias signal
20 and a magnetic field intercepting the Hall-effect element; and
 a component, which is accessible to a user from outside of the package, arranged with
 respect to the Hall-effect element such that a characteristic of the component may be changed
 to adjust a sensitivity of a gain of the Hall-effect device to temperature changes in the device.

25 10. The Hall-effect device as claimed in claim 9, wherein:
 the device further comprises a temperature-variable signal source arranged to provide
 the bias signal such that a value of the bias signal changes in response to temperature changes
 in the Hall-effect element, and
 the component is arranged with respect to the signal source such that the characteristic
30 of the component may be changed to adjust a sensitivity of the signal source to temperature
 changes in the Hall-effect element.

11. A Hall-effect device, comprising:
a substrate mounted in an integrated circuit package;
a Hall-effect element, integrated on the substrate, having at least one input for
receiving a bias signal and at least one output for providing an output signal dependent on the
bias signal and a magnetic field intercepting the Hall-effect element; and
means for adjusting a sensitivity of a gain of the Hall-effect device to changing
temperature conditions from a point external to the package.

12. The Hall-effect device according to claim 11, wherein the means for means for
adjusting the sensitivity of the gain of the Hall-effect device includes means for providing the
bias signal such that a value of the bias signal changes responsive to temperature changes in
the Hall-effect element.

13. A method for compensating for the effects of temperature variations on a Hall-
effect element, comprising the steps of:

(a) providing a monolithically integrated circuit including a Hall-effect device having
a gain that changes with changing temperature conditions; and

(b) coupling a component to a pair of external terminals of the integrated circuit to
adjust a sensitivity of the gain of the Hall-effect device to temperature variations.

14. The method of claim 13, further comprising the step of:

(c) altering a characteristic of the component to adjust the sensitivity of the gain of the
Hall-effect device to temperature variations.

15. The method of claim 13, wherein:

step (a) includes providing the monolithically integrated circuit including a Hall-effect
element and a temperature-variable signal source arranged to provide a bias signal to at least
one input of the Hall-effect element such that a value of the bias signal changes responsive to
temperature changes in the Hall device; and

step (b) includes coupling the component to the pair of external terminals of the
integrated circuit to adjust a sensitivity of the temperature-variable signal source to
temperature variations.

16. The method of claim 15, further comprising the step of:
- (c) altering a characteristic of the component to adjust the sensitivity of the temperature-variable signal source to temperature variations.

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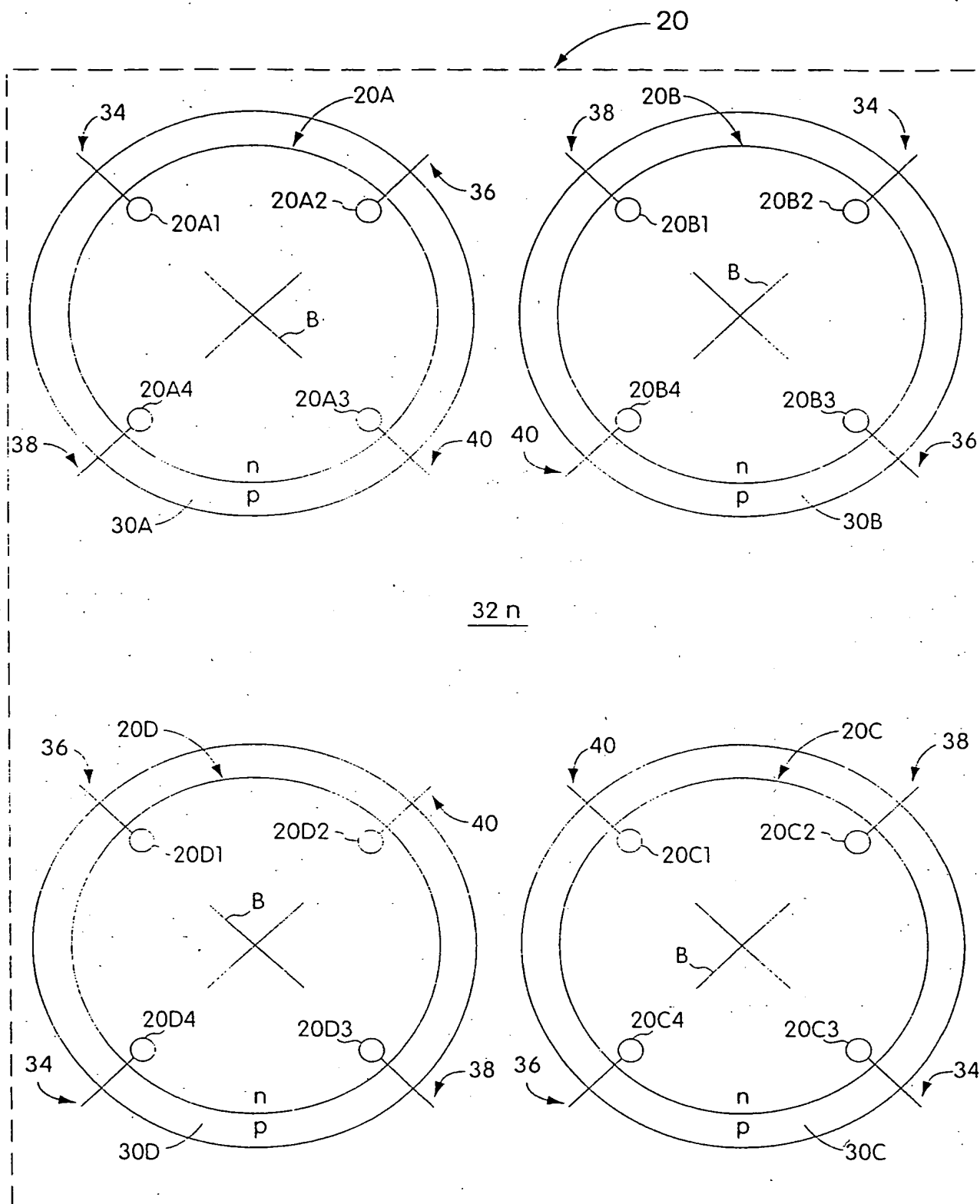


Fig. 1

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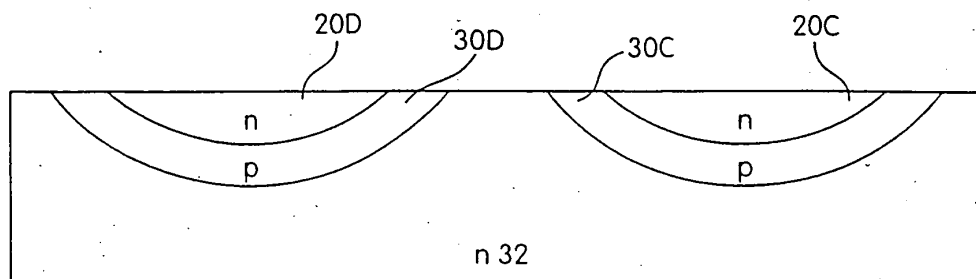


Fig. 2

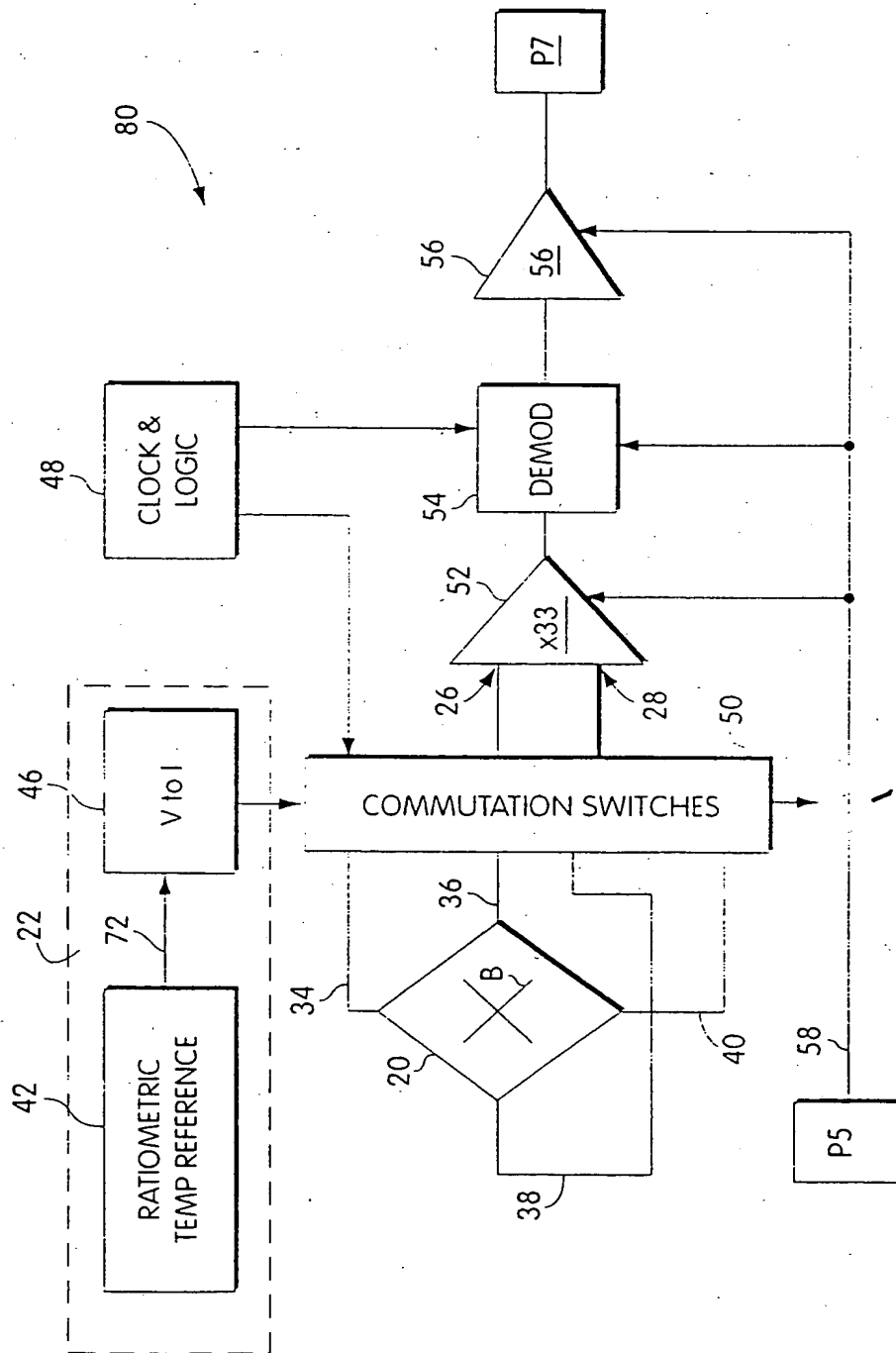


Fig.3

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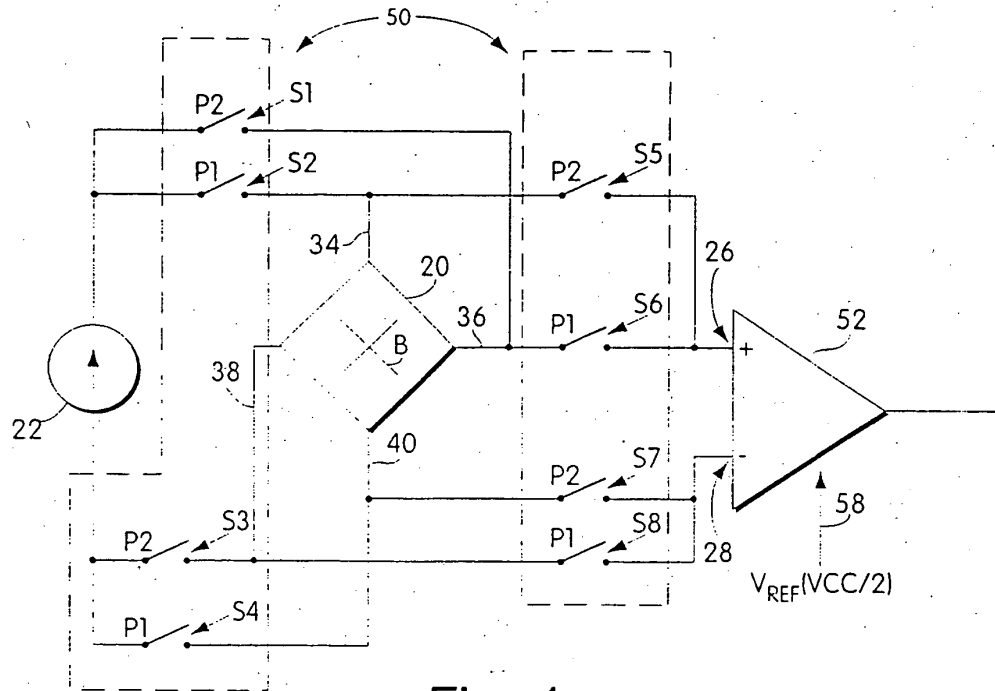


Fig. 4

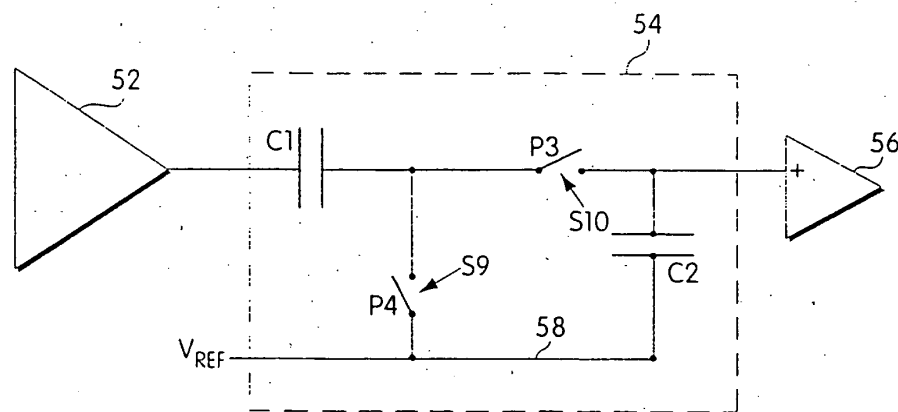


Fig. 5

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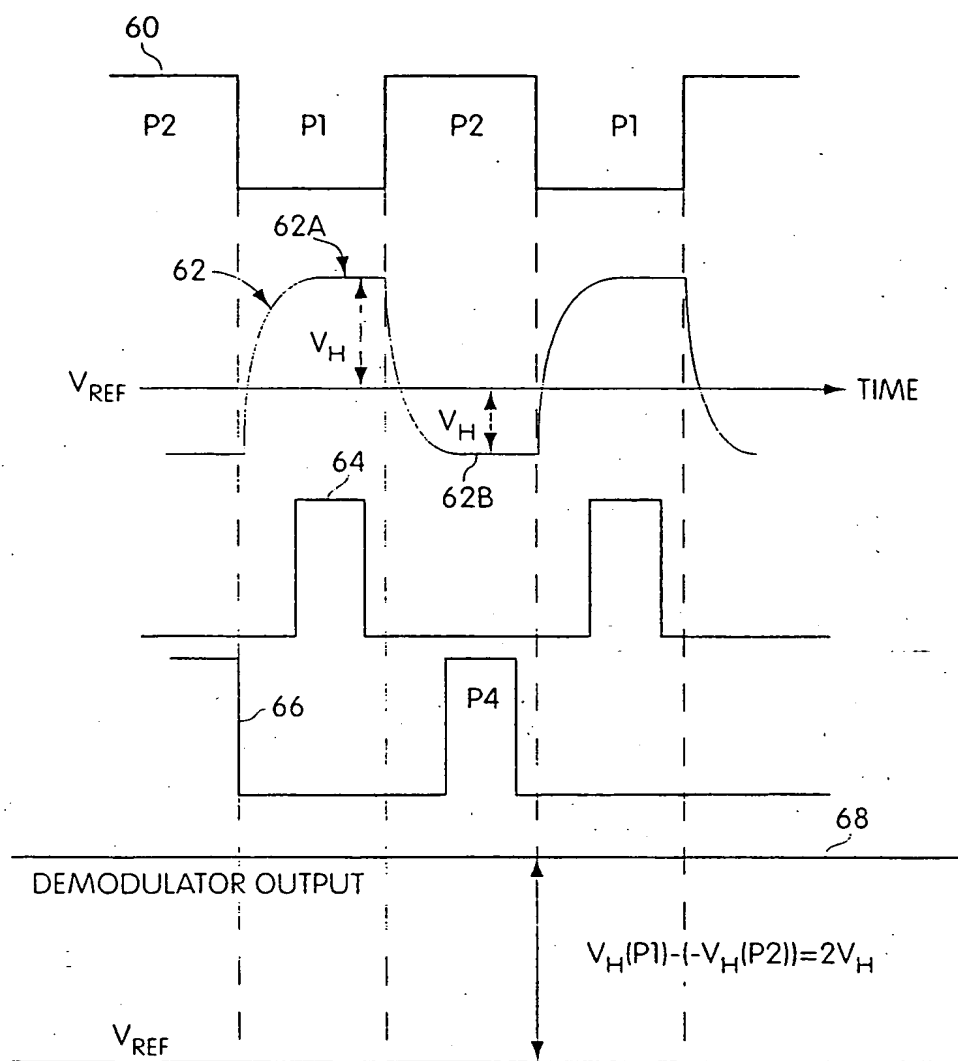


Fig. 6

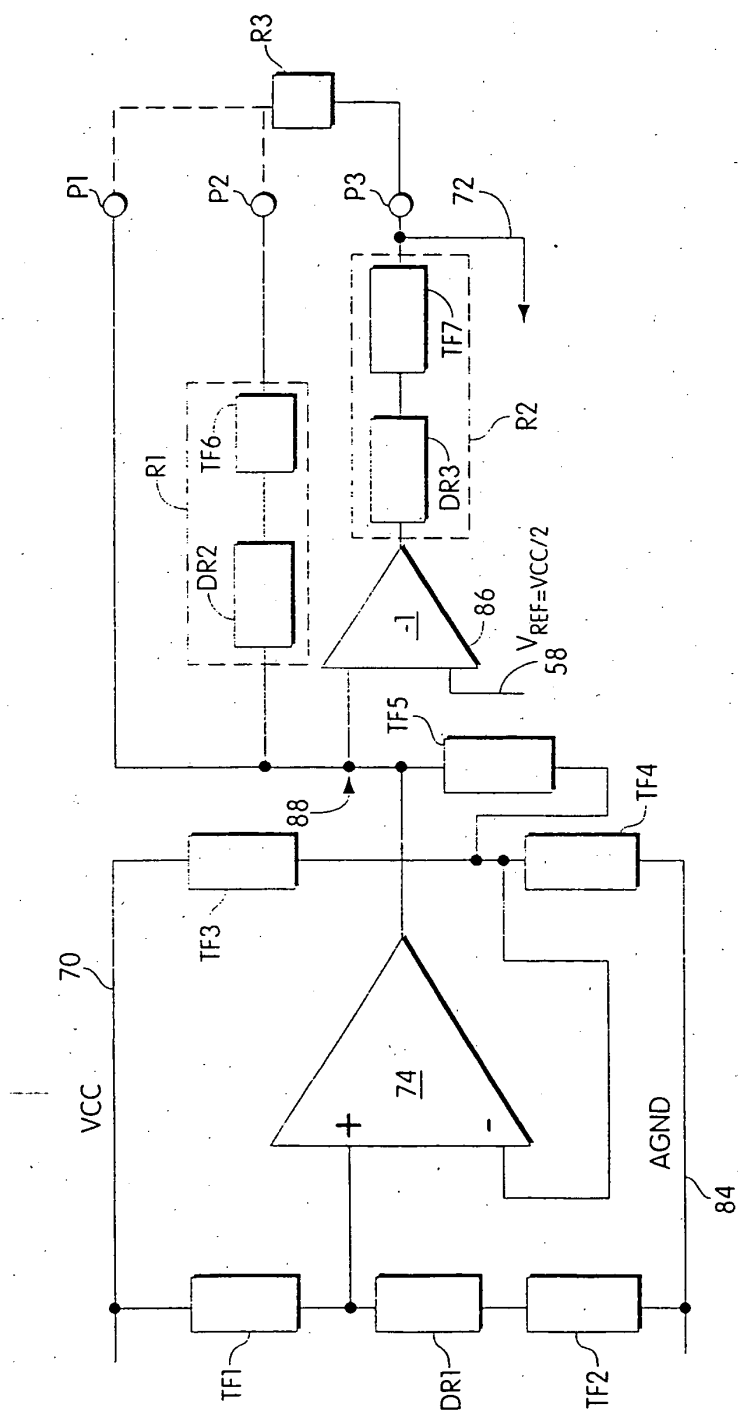


Fig. 7

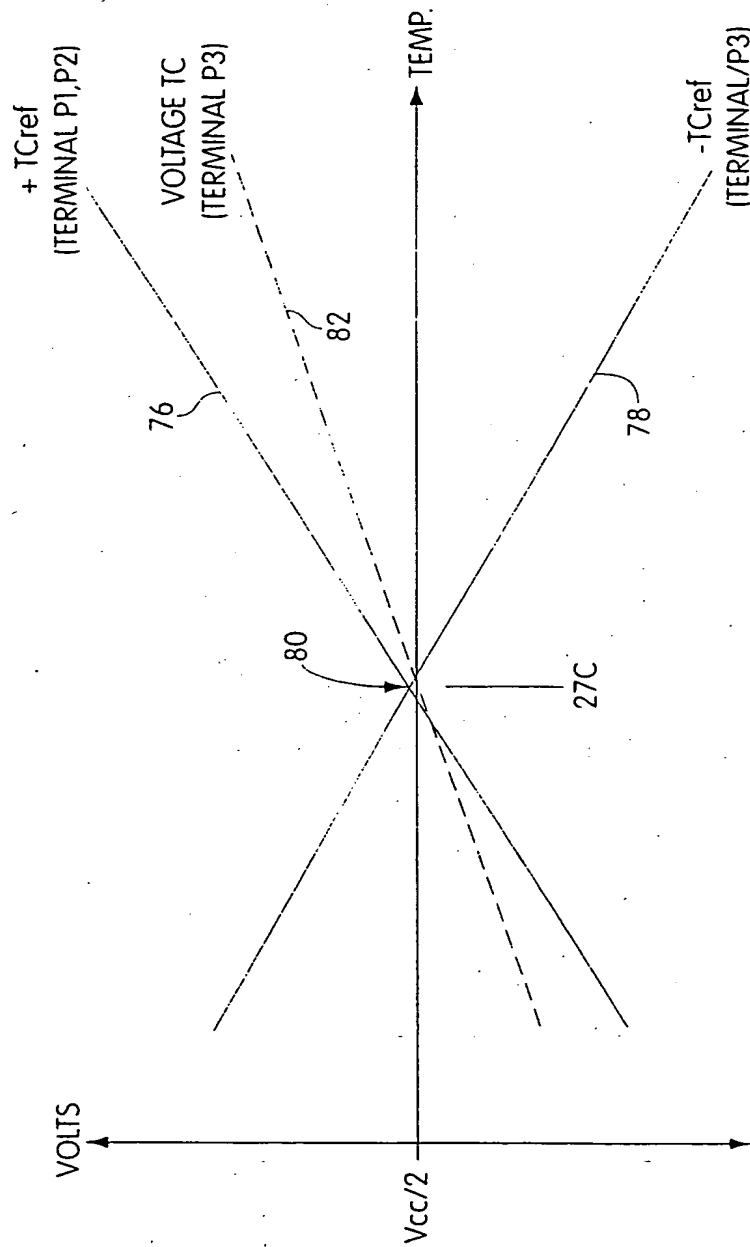


Fig. 8

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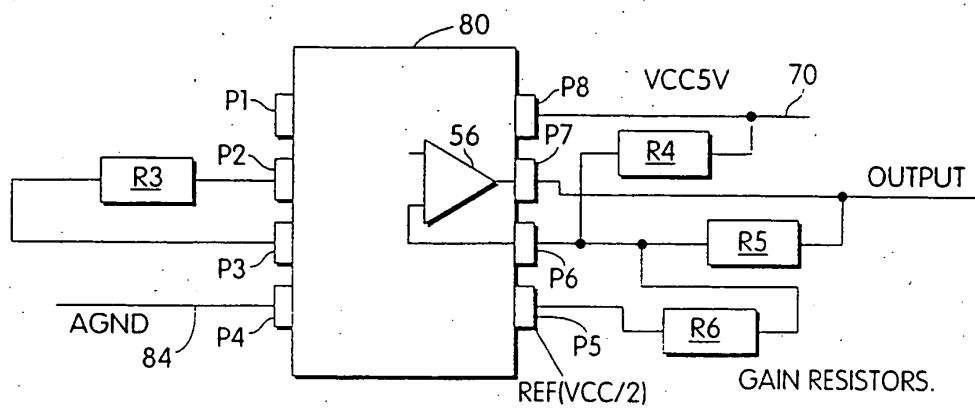


Fig. 9

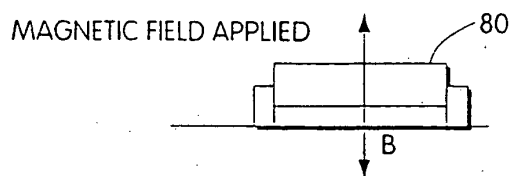


Fig. 10

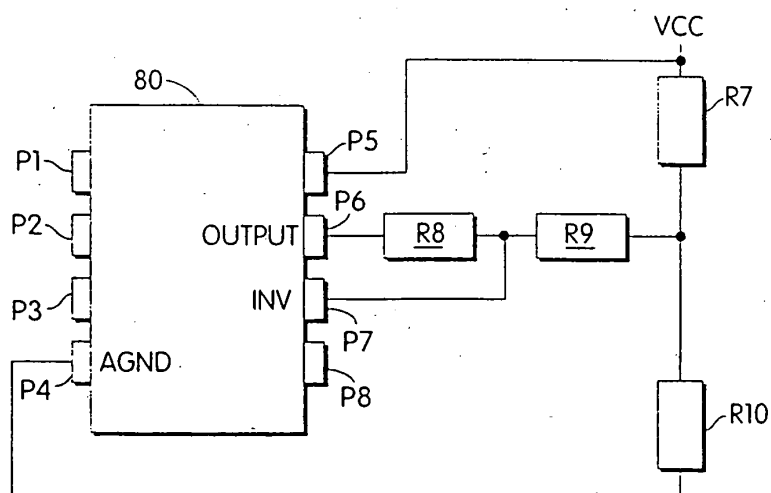


Fig. 11

INTERNATIONAL SEARCH REPORT

International Application No

PCT/US 98/16931

A. CLASSIFICATION OF SUBJECT MATTER

IPC 6 G01R33/07

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 6 G01R

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	US 5 604 433 A (THEUS ULRICH ET AL) 18 February 1997 see column 4, line 47 - line 64	1-3,8-16
Y	----	5-7
Y	DE 38 27 606 A (KOSTAL LEOPOLD GMBH & CO KG) 2 March 1989 see column 3, line 12 - line 57	5-7
X	US 4 734 594 A (NELSON RICHARD W) 29 March 1988 see column 2, line 7 - line 39	1-3, 9-13,15
X	GB 2 310 046 A (EATON CORP) 13 August 1997 see page 3, line 19 - page 5, line 24	1-4,9-13
A	US 5 389 889 A (TOWNE JAY M ET AL) 14 February 1995 -----	

☐ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

* Special categories of cited documents :

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Date of the actual completion of the international search

1 February 1999

Date of mailing of the international search report

08/02/1999

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Authorized officer

Swartjes, H

INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

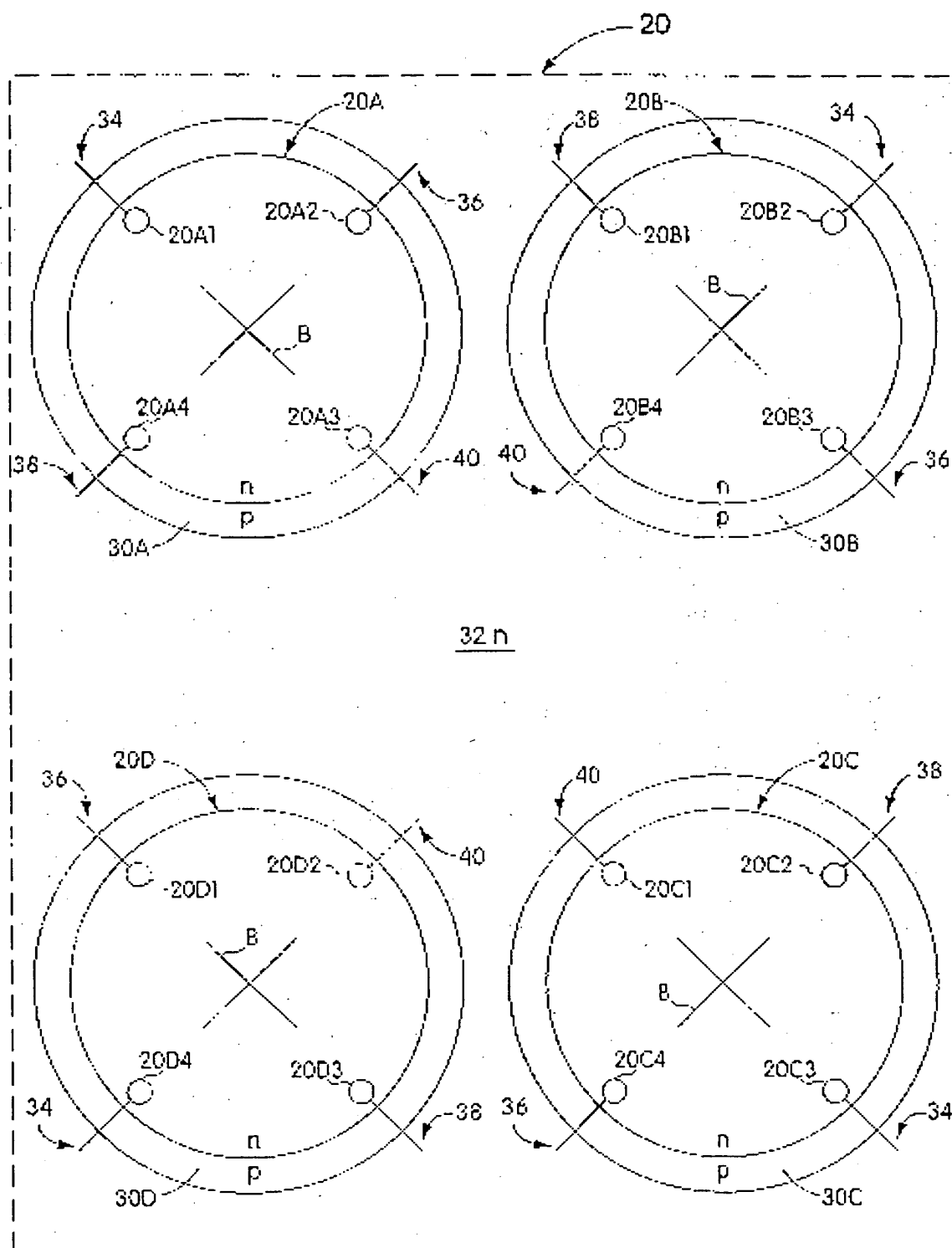
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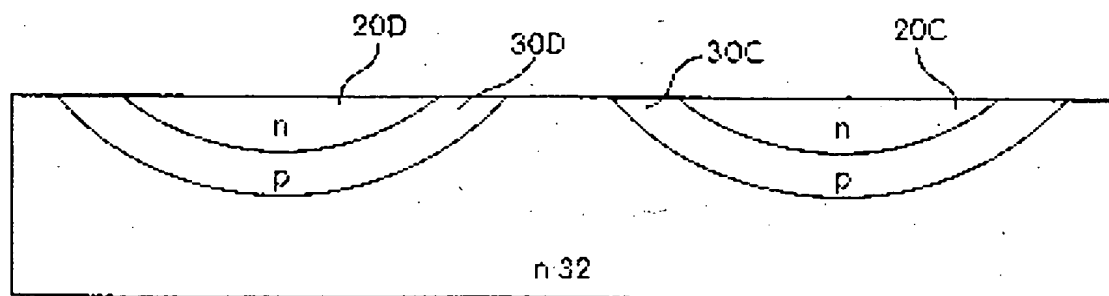
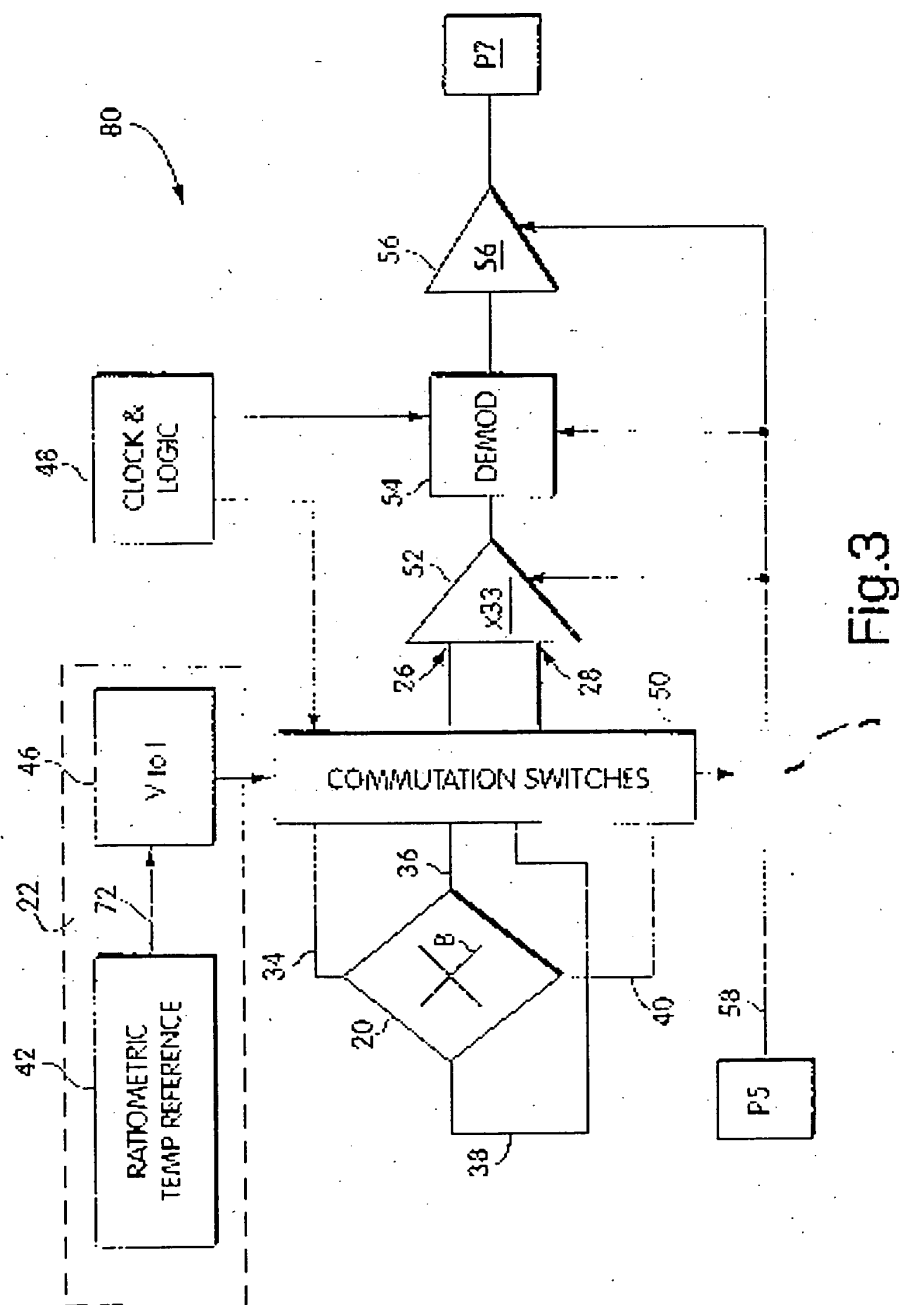


Fig. 2



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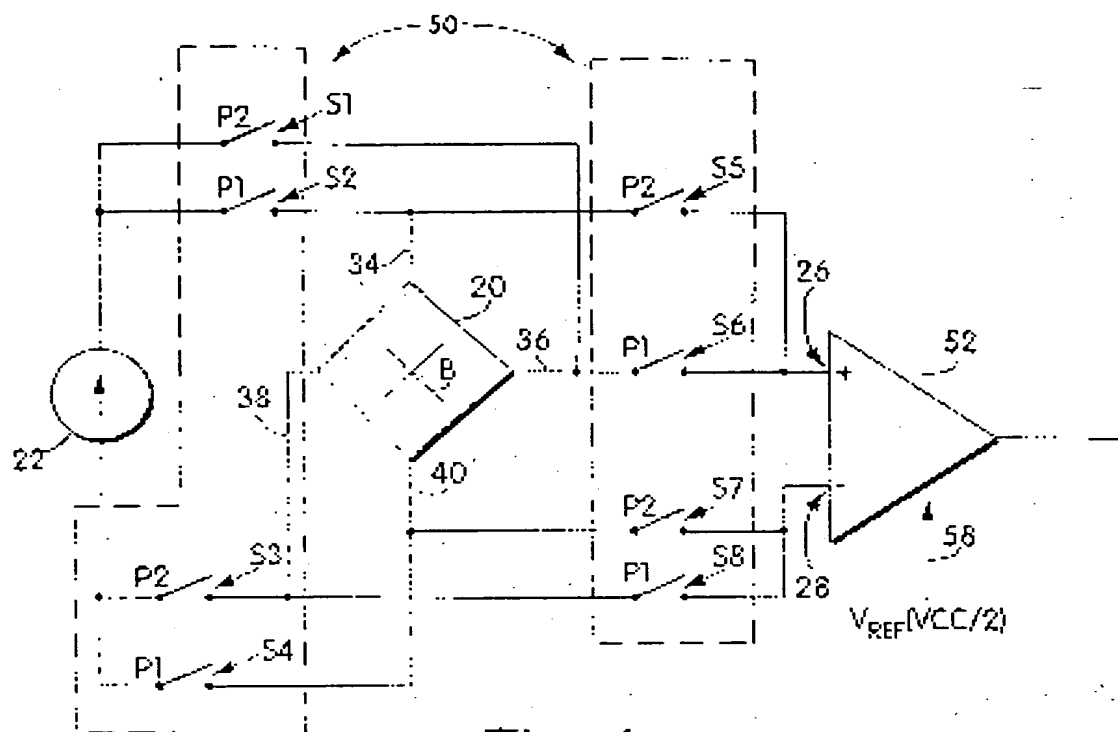


Fig. 4

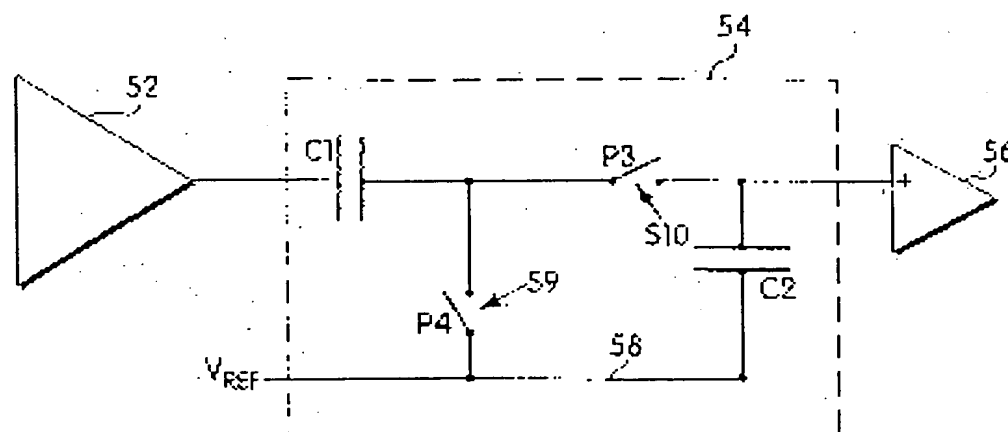


Fig. 5

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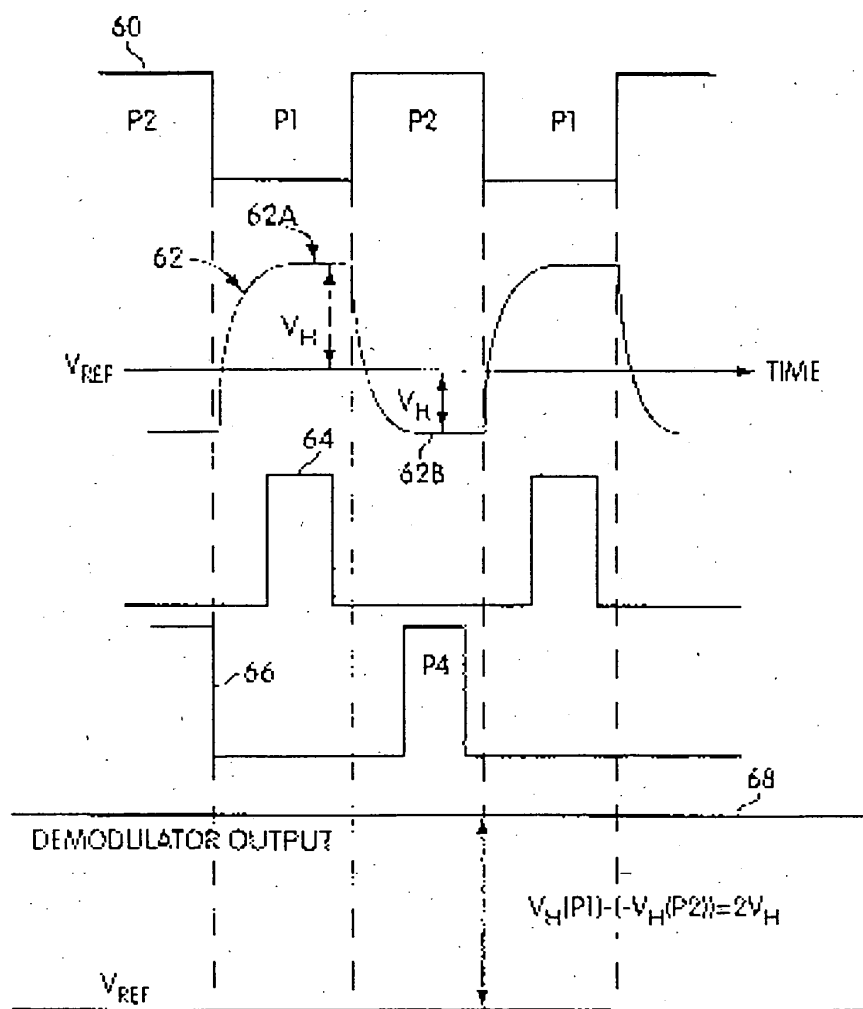
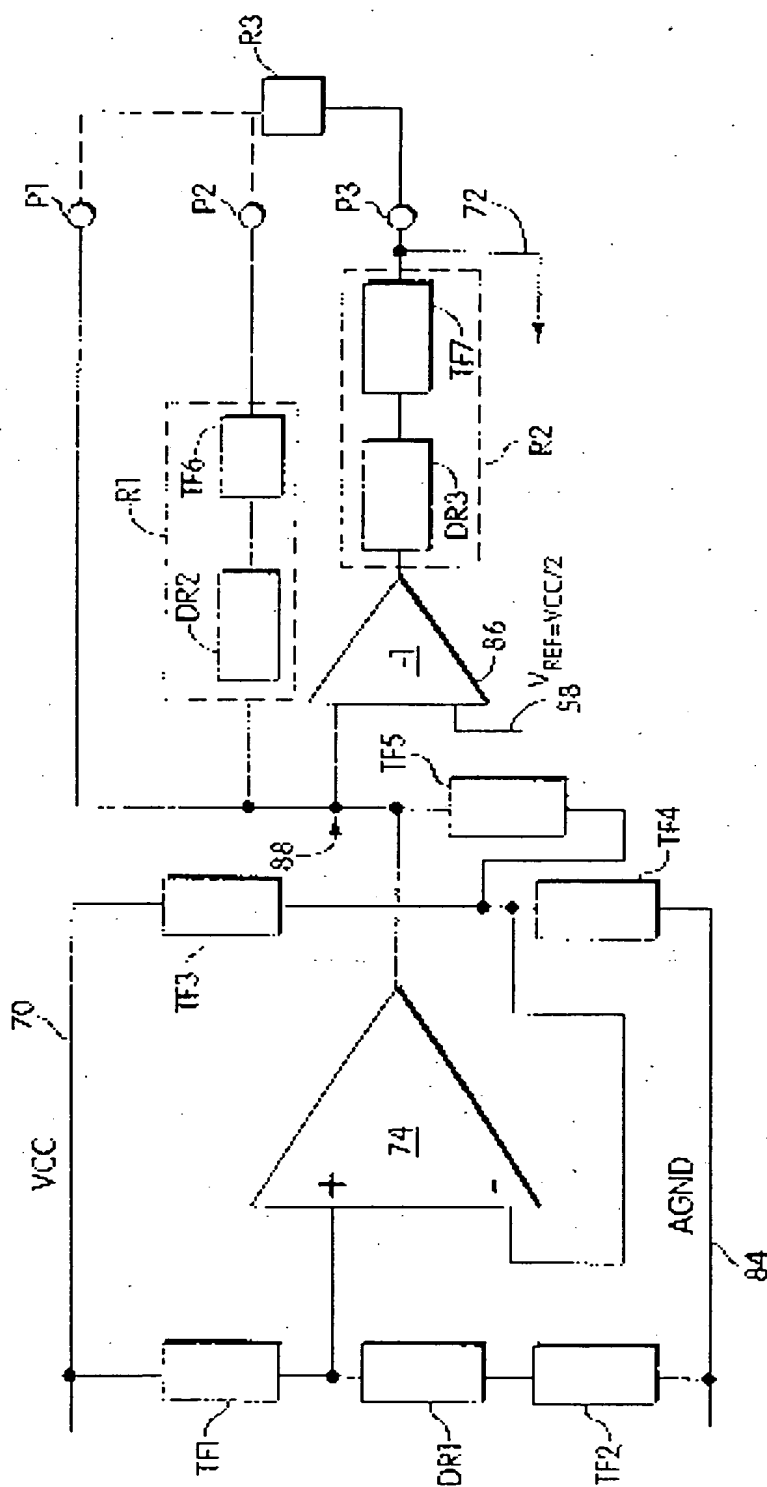


Fig. 6



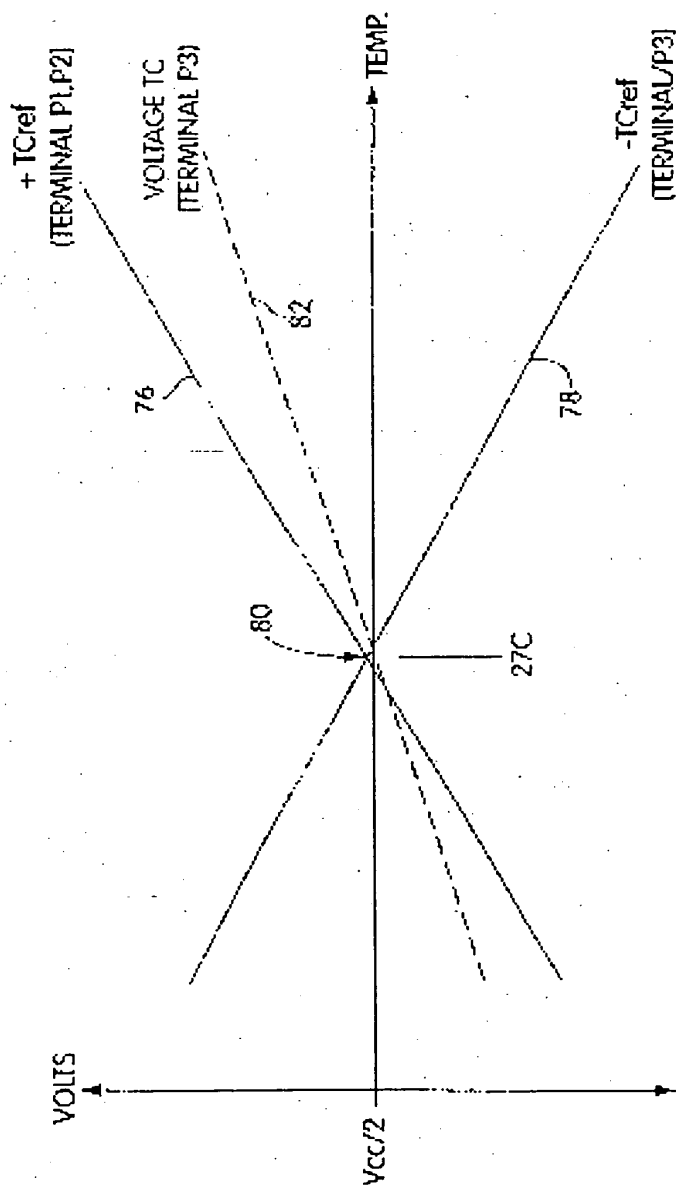


Fig. 8

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